# Interconnect Modeling using a Surface Admittance Operator Derived with the Fokas Method

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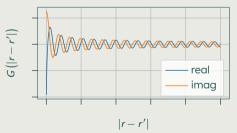


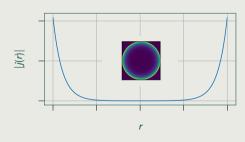
Quantum Mechanical & Electromagnetic Systems Modelling Lab



#### Context

- ► Trends in present-day electronic devices
  - ► Increasing operating frequencies
  - Continuing miniaturization
- Accurate full-wave modeling tools are required
- ▶ Skin and proximity effects in modern interconnects are challenging
  - ► Green's function integrals in surface-based methods
  - ► Fine discretization in volumetric methods





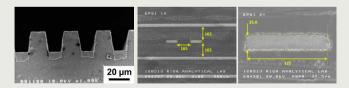




#### Context

#### **Differential surface admittance operator**<sup>1</sup>: exact and global boundary condition

- Current implementations
  - ► Limited to canonical shapes²
  - or reintroduction of conductive medium Green's function<sup>3</sup>
- Proposed: construction invoking Fokas method
  - ---- arbitrary homogeneous material properties and **polygonal shapes**
  - 1. D. De Zutter and L. Knockaert, "Skin effect modeling based on a differential surface admittance operator," IEEE Trans. Microw. Theory Techn., 2005
- T. Demeester and D. De Zutter, "Construction of the Dirichlet to Neumann boundary operator for triangles and applications in the analysis of polygonal conductors," IEEE Trans. Microw. Theory Techn., 2010
- U. R. Patel and P. Triverio, "Skin effect modeling in conductors of arbitrary shape through a surface admittance operator and the contour integral method," IEEE Trans. Microw. Theory Techn., 2016







Construction of the differential surface admittance (DSA) operator

Original situation

Equivalent situation

$$-\jmath\omega\mu_{\mathbf{i}}(\hat{\mathbf{n}}\times\mathbf{h}_{\mathbf{i}}) = \chi \mathbf{e}$$

$$\zeta_{m+1}$$

$$\zeta_{m}$$

Dirichlet-to-Neumann (DtN) operator  $\mathcal{X}'$  mapping e to its normal derivative

**Proposed:** new approach to construct the DSA operator



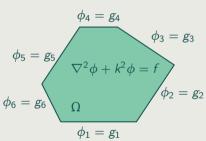


The Fokas method: general concept

Solution of **boundary value problem** (BVP) on n-sided polygon  $\Omega$ 

- ▶ Governed by partial differential equation (e.g., Helmholtz with real k)
- ▶ Subject to conditions on the sides  $\partial\Omega_j$  (e.g., Dirichlet)
- ► Find unknown boundary values (e.g., Neumann)

$$\begin{cases} \nabla^2 \phi + k^2 \phi = f & \text{in } \Omega, \ k \in \mathbb{R} \\ \phi_j = g_j & \text{on } \partial \Omega_j, \ j \in \{1, \dots, n\} \end{cases}$$
$$\frac{\partial \phi_j}{\partial n}?$$



A. S. Fokas, "A unified transform method for solving linear and certain nonlinear PDEs," Proc. Royal Soc. A, 1997





The Fokas method: interconnect modeling

- **Novel application** of **extended** Fokas method: determine  $\mathcal{X}$  and  $\mathcal{X}'$ 
  - ► Relation between  $\phi = e_z$  and its normal derivative  $\frac{\partial \phi}{\partial n} = -j\omega\mu_{(i,e)}h_{tan}^{(i,e)}$ ►  $e_z$  satisfies Helmholtz equations  $\nabla^2 e_z + k_{(i,e)}^2 e_z = 0$  with  $k_{(i,e)} \in \mathbb{C}$
- ► Reformulate well-known BVP using Fokas method's **global relation**

$$F(\lambda) = \int_{\mathcal{C}} \exp\left[-\frac{jk}{2}\left(\frac{\tilde{\zeta}}{\lambda} + \lambda \zeta\right)\right] \times \left[\frac{k\phi}{2}\left(\lambda \,\mathrm{d}\zeta - \frac{\mathrm{d}\tilde{\zeta}}{\lambda}\right) + \frac{\partial \phi}{\partial n}\,\mathrm{d}c\right] = 0, \ \forall \lambda \in \mathbb{C}$$

 $\zeta$ : position on C $\lambda$ : collocation points

 $\blacktriangleright$  Cast  $\phi$  on basis of P Legendre polynomials and evaluate F in L well-chosen  $\lambda$ 's





Extraction of resistance and inductance per unit of length

- ►  $L \gg P$ : overdetermined system in terms of unknown coefficients for  $\frac{\partial \phi}{\partial n}$
- ► Solution and projection on pulse basis yields discretized equation

$$\mathbf{j}_{s} = \left(\frac{\mathcal{X}'}{\jmath \omega \mu_{e}} - \frac{\mathcal{X}}{\jmath \omega \mu_{i}}\right) \mathbf{e} \triangleq \mathcal{Y} \mathbf{e} \qquad \longrightarrow \qquad \overline{\overline{G}} \mathbf{J} = \left(\frac{\overline{\overline{X}'}}{\jmath \omega \mu_{e}} - \frac{\overline{\overline{X}}}{\jmath \omega \mu_{i}}\right) \mathbf{E} \triangleq \overline{\overline{Y}} \mathbf{E}$$

lacktriangle Extract resistance and inductance matrices  $\overline{\overline{R}}$  and  $\overline{\overline{L}}$ 

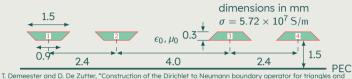
$$\overline{\overline{R}} + j\omega \overline{\overline{L}} = \left(\overline{\overline{T}}^{\mathsf{T}} \left(\overline{\overline{\overline{G}}} \overline{\overline{Y}}^{-1} \overline{\overline{\overline{G}}} + j\omega \overline{\overline{A}}\right)^{-1} \overline{\overline{T}}\right)^{-1}$$

with  $\overline{\overline{G}}$  a Gramian matrix,  $\overline{\overline{A}}$  the interaction matrix and  $\overline{\overline{T}}$  an incidence matrix





Asymmetric configuration with four trapezoidal conductors



applications in the analysis of polygonal conductors," IEEE Trans. Microw. Theory Techn., 2010

#### Improvements w.r.t. the reference

- No need to model trapezoid as a composition of a rectangle and two triangles
- ► No tweaking required to eliminate Gibbs effect in triangles



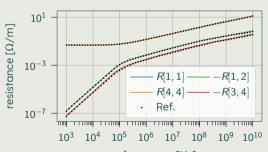


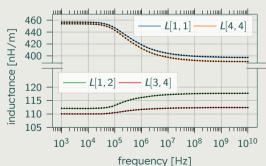


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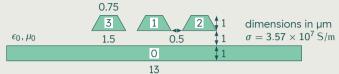




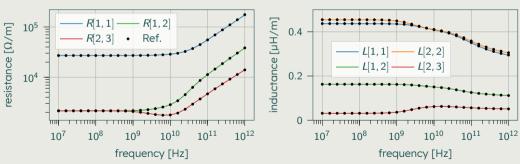
mec

frequency [Hz]

Multiconductor transmission line with finite reference plane



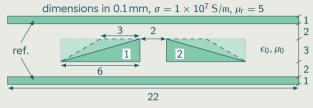
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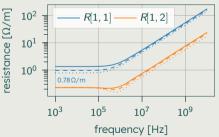


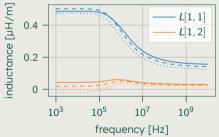
Influence of the conductor shape



Pouillet resistances:

$$R_{11}^{DC} = \frac{1}{\sigma} \left( \frac{1}{A_{\text{sig}}} + \frac{1}{A_{\text{ref}}} \right)$$
$$R_{12}^{DC} = \frac{1}{\sigma A_{\text{ref}}}$$









## Conclusion

Novel interconnect modeling framework

- Robust construction of differential surface admittance operator for arbitrary polygonal cross-sections
- Extension of Fokas method to complex eigenvalues to solve pertinent partial differential equations
- ► Efficient and accurate extraction of **per-unit-of-length resistance and inductance** parameters





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